

# Ch3: Semiconductor Electronics – Part 1

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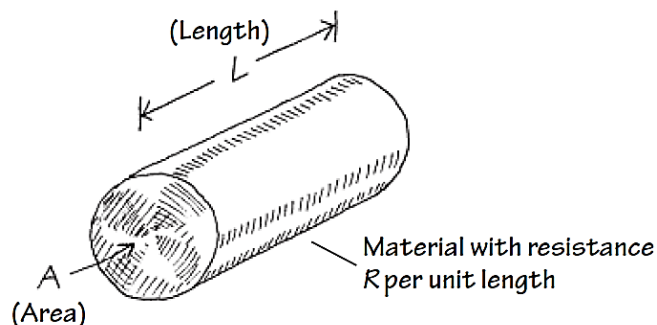
# Semiconductor Physics



# Semiconductors

Materials classification by their **ability to conduct electricity**:

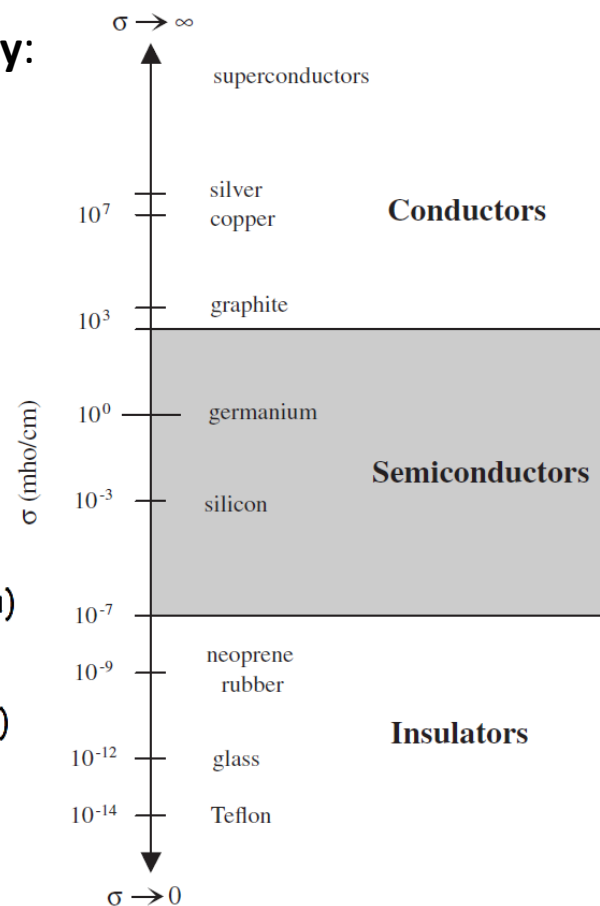
- **Conductors** that easily pass an electric current, such as silver and copper.
- **Insulators** that do not pass an electric current, such as rubber, wood, and glass.
- **Semiconductor** whose conductivity lies between those of conductors and insulators, such as silicon and germanium.



$$\rho = R \frac{A}{L} \quad (\text{Resistivity ohm}\cdot\text{cm})$$

$$\sigma = \frac{1}{\rho} \quad (\text{Conductivity mho/cm})$$

$$\text{mho} = \frac{1}{\text{ohm}} = \frac{1}{\Omega} = \mathcal{U}$$



There are two types of **semiconductors** based on their structure: **Intrinsic** and **Extrinsic**.

# Intrinsic Semiconductors

**Intrinsic Semiconductors** are **Pure** form of semiconductors like **Silicon (Si)** which is the most important semiconductor used in building electrical devices. In pure form, conductivity of silicon depends on **temperature**.

Group classification: IV

Atomic number: 14

Symbol: Si

Atomic weight: 28.086

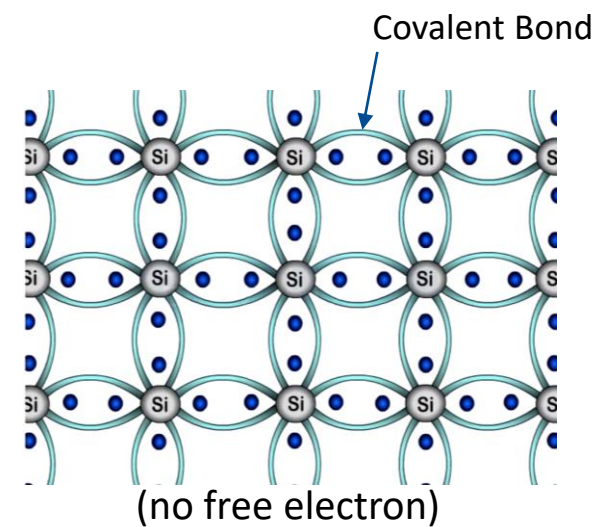
Name: Silicon

Orbitals:  $1s^2, 2s^2, 2p^6, 3s^2, 3p^2$

4 valence electrons

**4** valence electrons orbiting in its outermost shell. The atom needs a total of **8** electron in its outer shell to become **stable**. So, silicon atoms are bonded together and form the structure.

1 H																	2 He
3 Li	4 Be											5 B	6 C	7 N	8 O	9 F	10 Ne
11 Na	12 Mg											13 Al	14 Si	15 P	16 S	17 Cl	18 Ar
19 K	20 Ca	21 Sc	22 Ti	23 V	24 Cr	25 Mn	26 Fe	27 Co	28 Ni	29 Cu	30 Zn	31 Ga	32 Ge	33 As	34 Se	35 Br	36 Kr
37 Rb	38 Sr	39 Y	40 Zr	41 Nb	42 Mo	43 Tc	44 Ru	45 Rh	46 Pd	47 Ag	48 Cd	49 In	50 Sn	51 Sb	52 Te	53 I	54 Xe





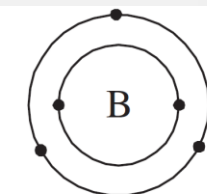
# Extrinsic Semiconductors

Intrinsic semiconductors act more as an insulator than a conductor (since they do not have free electrons), and they do not have the ability to change conductive states when an external voltage is applied. Hence, **impurities** should be added to intrinsic semiconductors **to improve conductivity** and make **Extrinsic Semiconductors**. This process is called **Doping**.

Many ingredients can be added in the doping process. Two of the most important ingredients that can alter the electrical conductivity dramatically are **Phosphorus (P)** and **Boron (B)**.

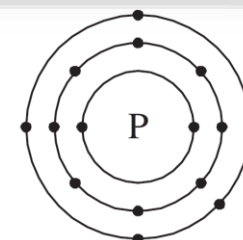
1 H																	2 He
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37 Rb	38 Sr	39 Y	40 Zr	41 Nb	42 Mo	43 Tc	44 Ru	45 Rh	46 Pd	47 Ag	48 Cd	49 In	50 Sn	51 Sb	52 Te	53 I	54 Xe

B	5
10.811	
Boron	



Atomic configuration

P	15
30.974	
Phosphorus	



Atomic configuration

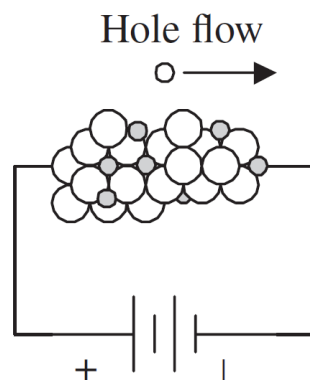




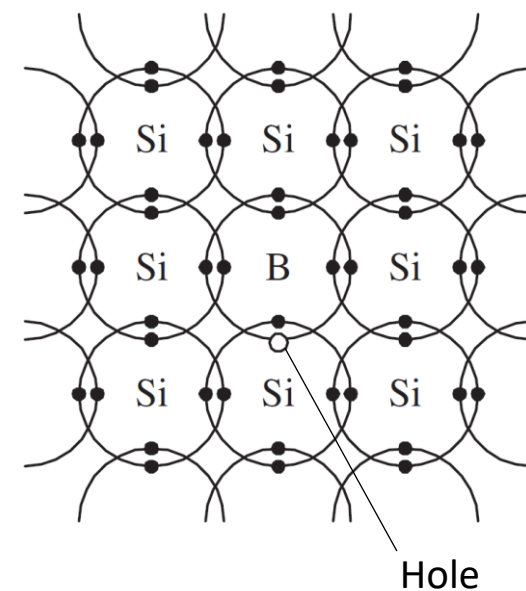
# Extrinsic Semiconductors (p-Type)

**Boron (B)** has three valence electrons. All three of its valence electrons will form covalent bonds with the valence electrons of neighboring silicon atoms and there will be a vacant spot (called a **hole**) within the covalent bond between one boron and one silicon atom.

If a voltage is applied across the silicon-boron mixture, the **hole** (which acts like a **positive charge!**) will move toward the **negative voltage end**, while a neighboring electron will fill in its place (**consequently, electrons flow**).



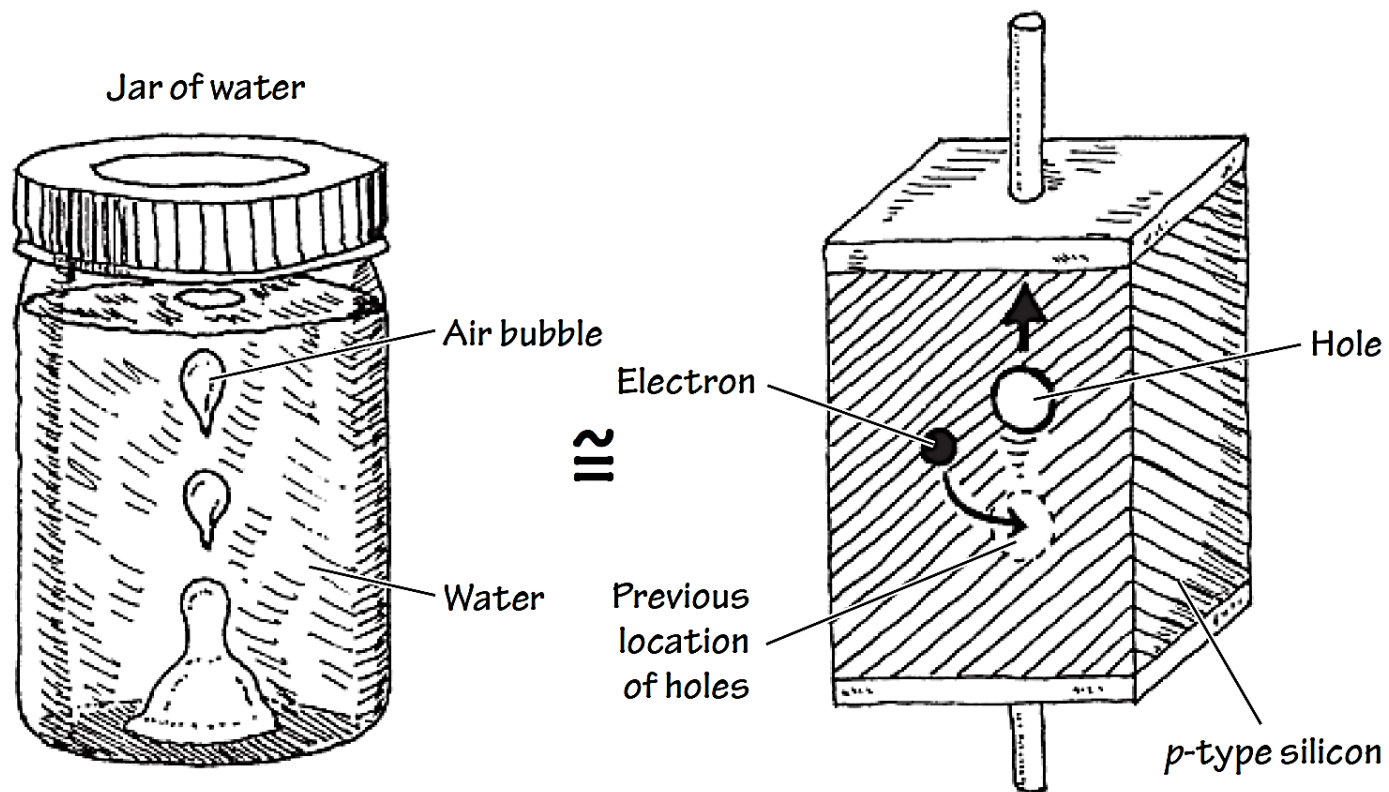
*p*-type silicon



Silicon that is doped with boron is referred to as **p-type silicon**, or **positive-charge-carrier-type silicon**.

# How Can Holes Move?

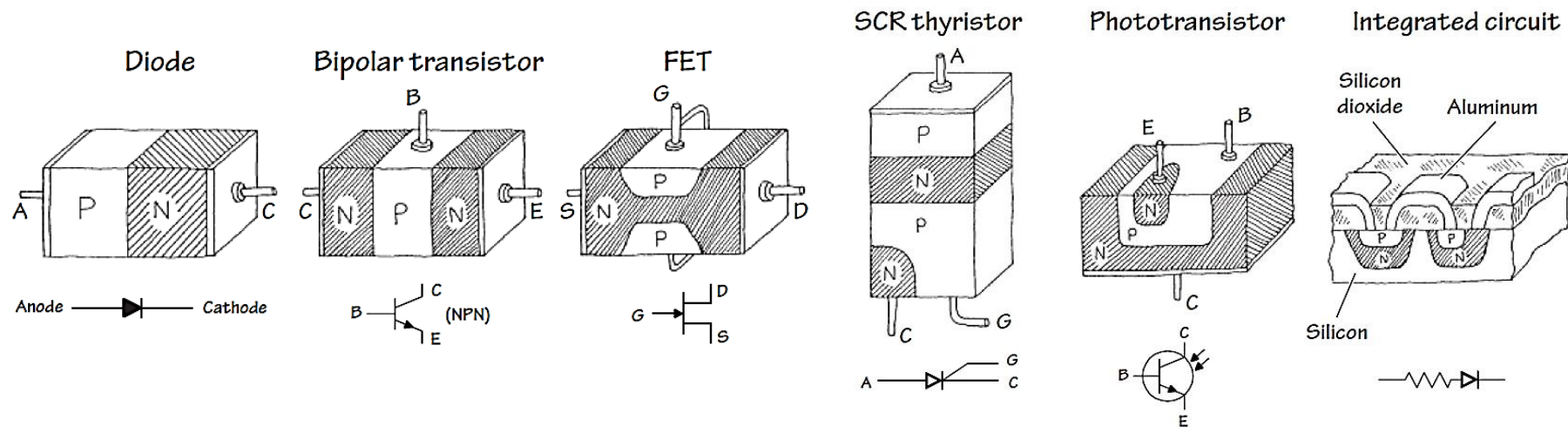
Turn a sealed bottle of water upside down and then right side up. The bubble trapped in the bottle will move in the opposite direction of the water!



# Applications

We now have two new conductors with two unique ways of passing an electric current, one does it with **extra unbound electrons (n-type silicon)**, the other does it with **holes (p-type silicon)**.

The **interaction** between n-type and p-type semiconductor materials in different configurations/combinations is the basis for most semiconductor electronic devices.



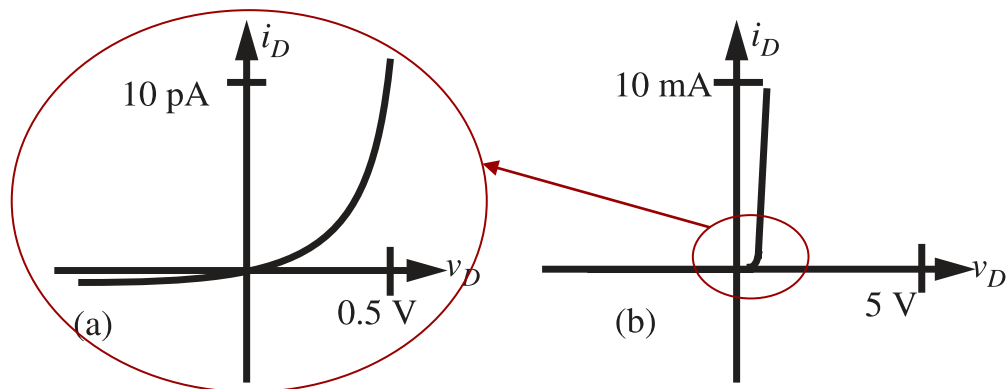
# Diodes



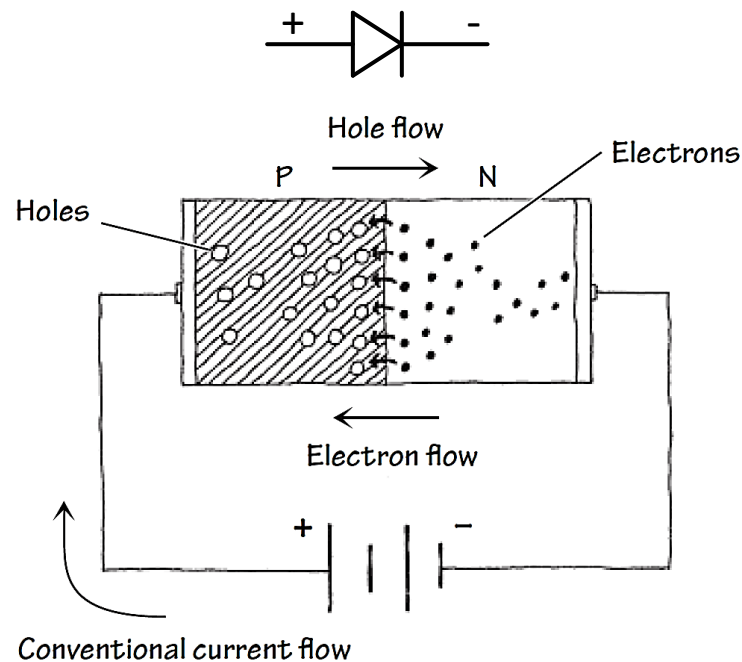


# Forward-Biased

When a diode is connected to a **voltage source** as shown, electrons from the n-type side and holes from the p-type side are forced toward the center (p-n interface). As the applied voltage approaches the value of the contact potential (0.6-0.7 V for silicon, 0.2-0.3 V for germanium), the **depletion region shrinks**, the electrons and holes combine, and the **current flow increases exponentially**.



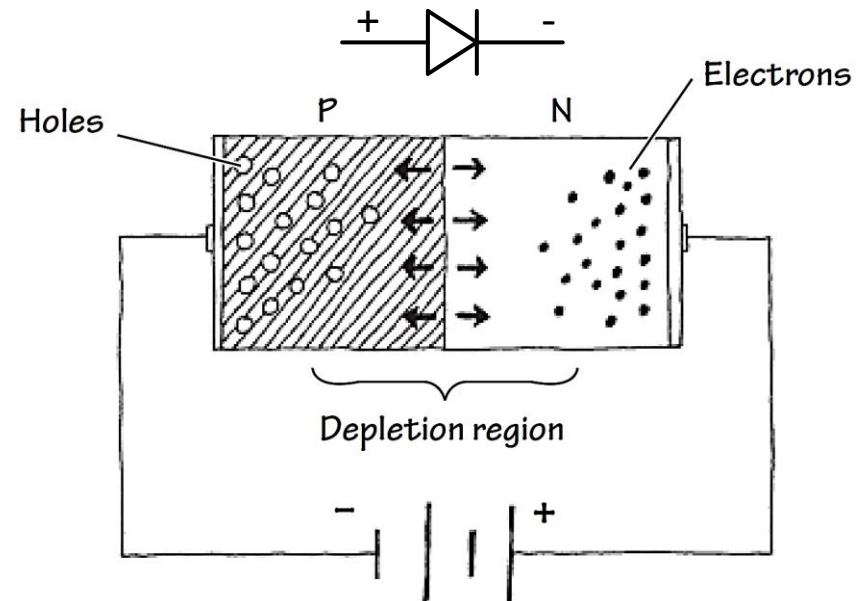
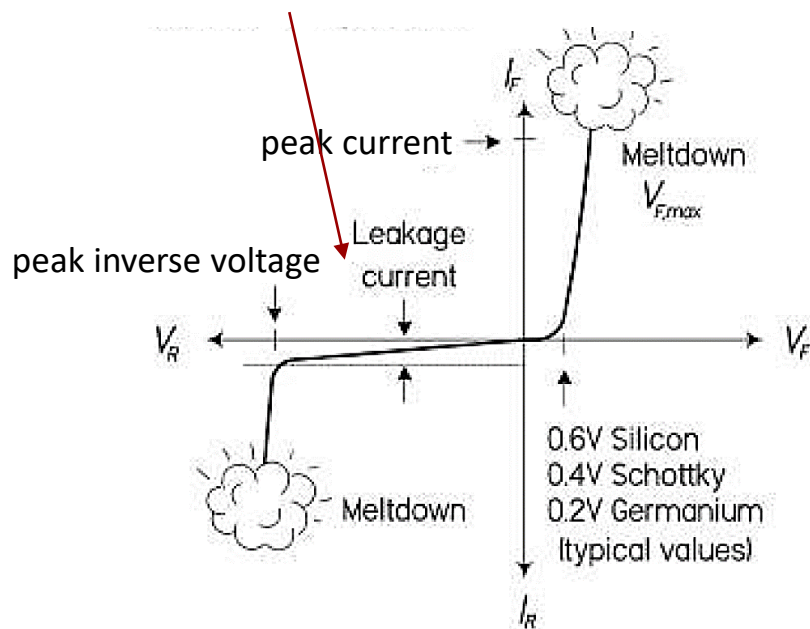
Hence, diode requires a specific voltage to **turn on**. Thus, they can also act as a voltage-sensitive switch.



# Reverse-Biased

When a diode is connected to a **voltage source** as shown, holes in the n-type side are forced to the left, while electrons in the p-type side are forced to the right. Consequently, the **depletion region enlarges** and **prevents the current flows** through the diode.

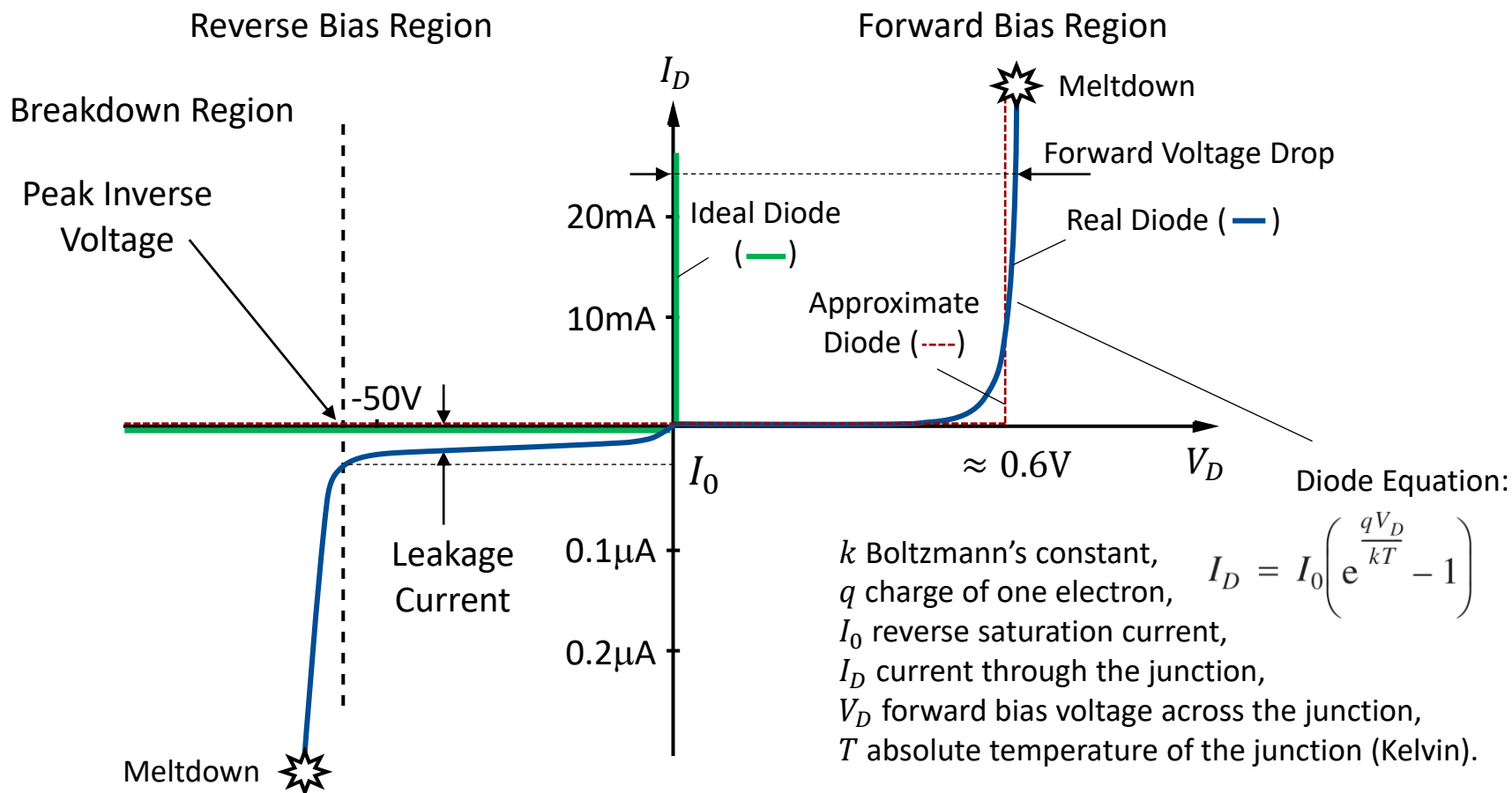
A reverse saturation current ( $I_0$ ) does flow, but it is **extremely small** ( $10^{-9}$  to  $10^{-15}$  A!).







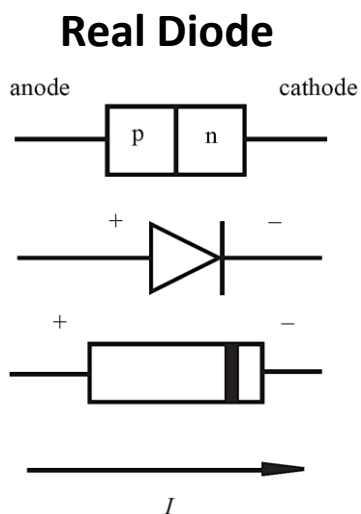
# Ideal, Approximate, and Real Diode



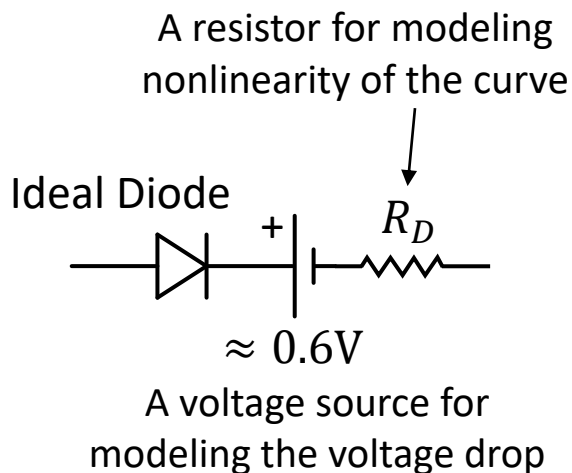
An **ideal diode** has **zero resistance** (short circuit) **when forward biased** and **infinite resistance** (open circuit) **when reverse biased**.

# Diode Modeling

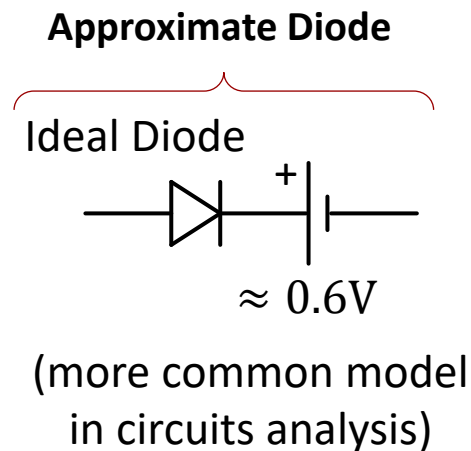
- Modeling of a diode in circuits analysis:



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- The heat generated in a diode:

$$P = V_D I_D$$

(Heat)

# Selection of Diodes

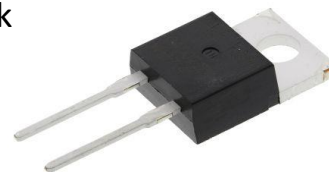
DEVICE	TYPE	PEAK INVERSE VOLTAGE PIV (V)	MAX. FORWARD CURRENT $I_{O(MAX)}$	MAX. REVERSE CURRENT $I_{R(MAX)}$	PEAK SURGE CURRENT $I_{FSM}$	MAX. VOLTAGE DROP $V_F(V)$
1N914	Fast Switch	90	75 mA	25 nA		0.8
1N4148	Signal	75	10 mA	25 nA	450 mA	1.0
1N4445	Signal	100	100 mA	50 nA		1.0
1N4001	Rectifier	50	1 A	0.03 mA	30 A	1.1
<b>1N4002</b>	Rectifier	100	1 A	0.03 mA	30 A	1.1
1N4003	Rectifier	200	1 A	0.03 mA	30 A	1.1
1N4004	Rectifier	400	1 A	0.03 mA	30 A	1.1

- Diodes require **nanoseconds** to switch between their on and off states. This switching **time** is fast enough for most applications, but when designing high-speed circuits it may pose a constraint.

- Power diodes** are capable of carrying very large currents. They are designed to be attached to **heat sinks** in order to efficiently **dissipate heat** produced in the junction.

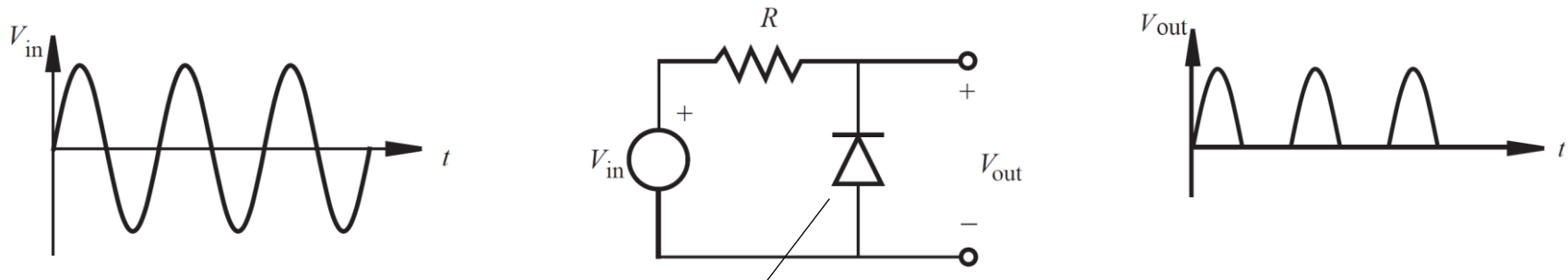


heat sink



# Application: Half-Wave Rectifier

**Half-Wave Rectifier** is used for passing only the positive (or negative) half of an AC signal. These circuits are used in the design of power supplies, where AC power must be transformed into DC power for use in electronic devices and digital circuits.



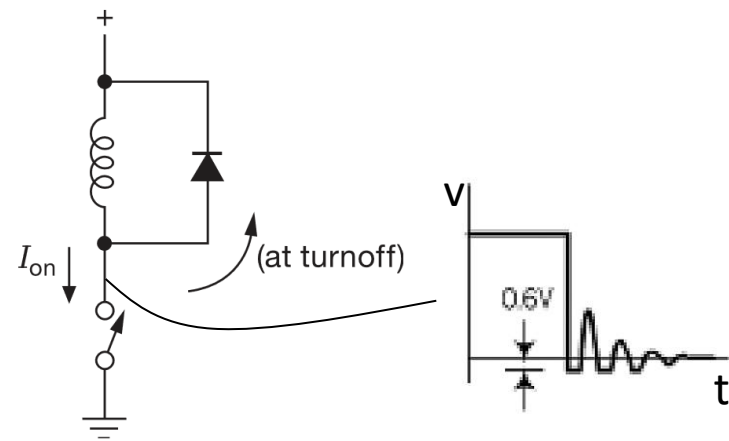
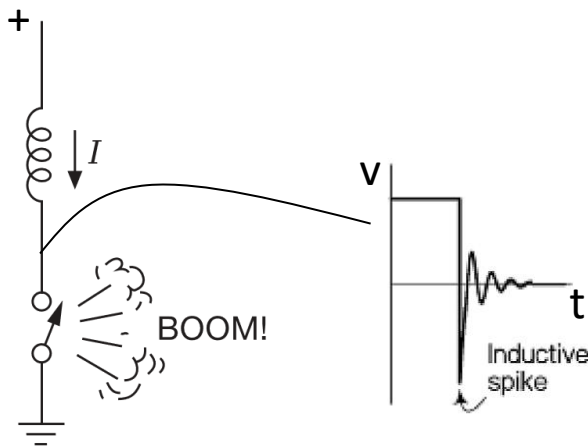
Assumption: **Ideal Diode**

When  $V_{in} > 0$ , the diode is reverse biased (open circuit), then output  $V_{out} = V_{in}$ .  
When  $V_{in} < 0$ , the diode is forward biased (short circuit), then  $V_{out} = 0$ .

❖ How the output would be if **Approximate Diode Curve** is assumed?

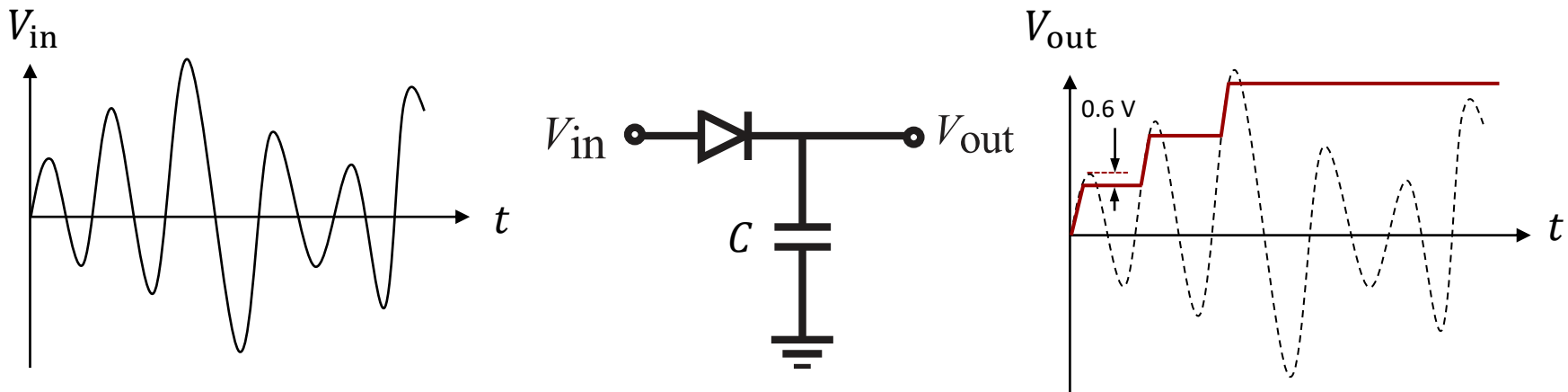
# Application: Fly-Back Diodes

When current flowing through an inductor (e.g., relay coils, motors, solenoids, ...) is **suddenly switched off**, the collapsing magnetic field will generate a **high-voltage spike** in the inductor's coils. This spike can damage the switch and can create electromagnetic interference (EMI) that can affect surrounding circuits. A diode placed across the relay's coil can protect neighboring circuitry by providing a short circuit for the high-voltage spike.



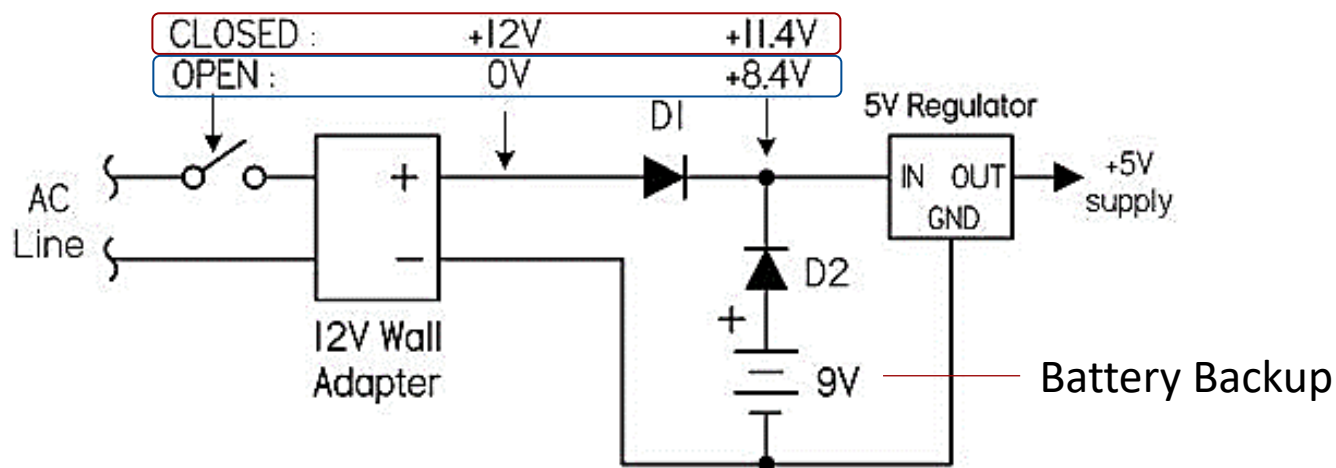
# Application: Peak Detector

The following circuit is known as a **peak detector**. When a time-varying signal  $V_{in}$  is applied at the input, the output signal  $V_{out}$  retains the maximum positive value of the input signal.



# Application: Simple Battery Backup

Devices are powered by a wall adapter with battery backup. Normally, if the switch is closed, power is delivered to the load from the 12 V wall adapter through  $D_1$  (note that  $D_2$  is reverse-biased, since its negative end is 2.4V more positive than its positive end). If power is interrupted (switch opened),  $D_1$  stops conducting, and the battery kicks in, sending current through  $D_2$  into the load ( $D_1$  blocks current from flowing back into the wall adapter).



**Drawback:** There is a penalty (0.6 V) for using diodes for battery backup. You can look for other solutions using transistors and ICs.



# Analysis of Diode Circuits

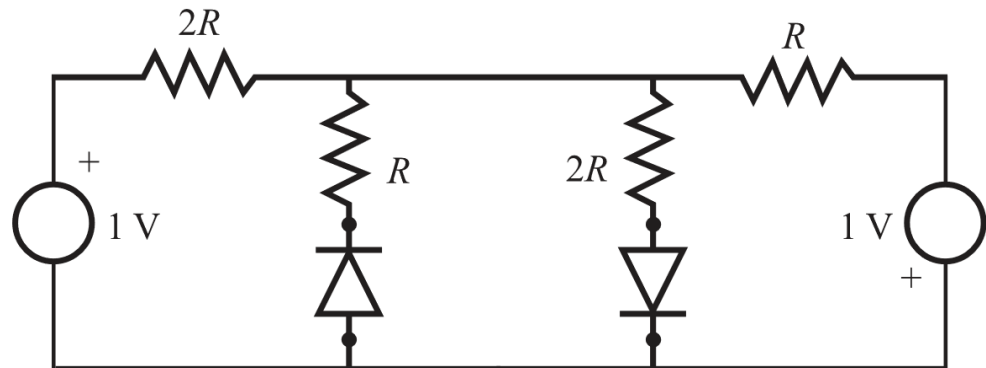
Since the **diode is a nonlinear device**, you can not naively apply the linear circuit analysis methods.

## Recommended method:

1. Assume current directions for **each circuit element**.
2. Replace each diode with an equivalent **open circuit** if the assumed current is in a reverse bias direction or a **short circuit** (or a small voltage source) if it is in the forward bias direction.
3. Compute the voltage drops and currents in the circuit loops using KVL and KCL.
4. If the **sign** on a resulting current is **opposite** to the assumed direction through an element, you have made the wrong assumption and must change its direction and **reanalyze** the circuit.
5. Repeat this procedure with **different combinations** of current directions until there are no inconsistencies between assumed and calculated voltage drops and currents.

# Example

In the following circuit, determine all currents and voltages. Assume that the diodes are ideals.



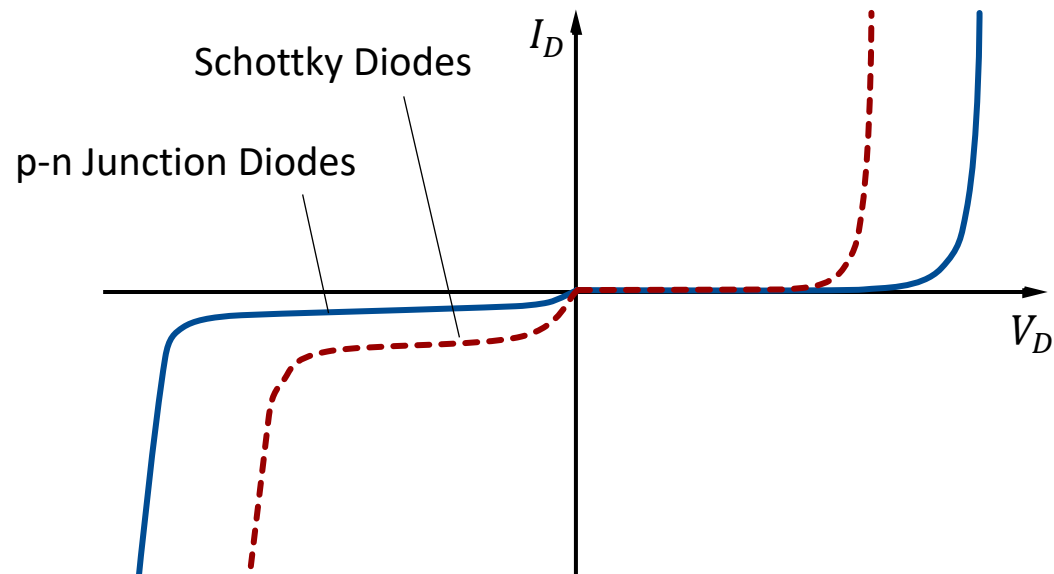
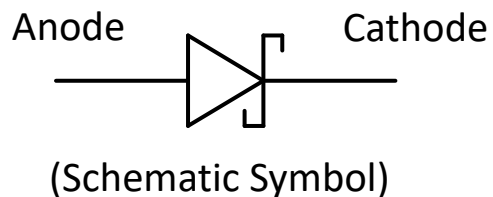
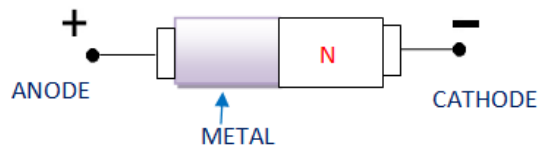
\* Now, determine all currents and voltages by using the approximate model of the diodes.

# Schottky Diodes

**Schottky Diodes** are similar in operation to p-n junction diodes but designed with **metal-semiconductor (n-type) junction** instead of a p-n junction. These diodes are used in similar applications as p-n junction diodes, but offer

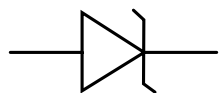
- Lower forward-bias voltage (0.15 to 0.45 V)  $\Rightarrow$  Lower power loss,
- Quicker switching times (useful in high-frequency applications).

**However**, they have lower reverse voltage rating and higher reverse leakage current.



# Zener Diode

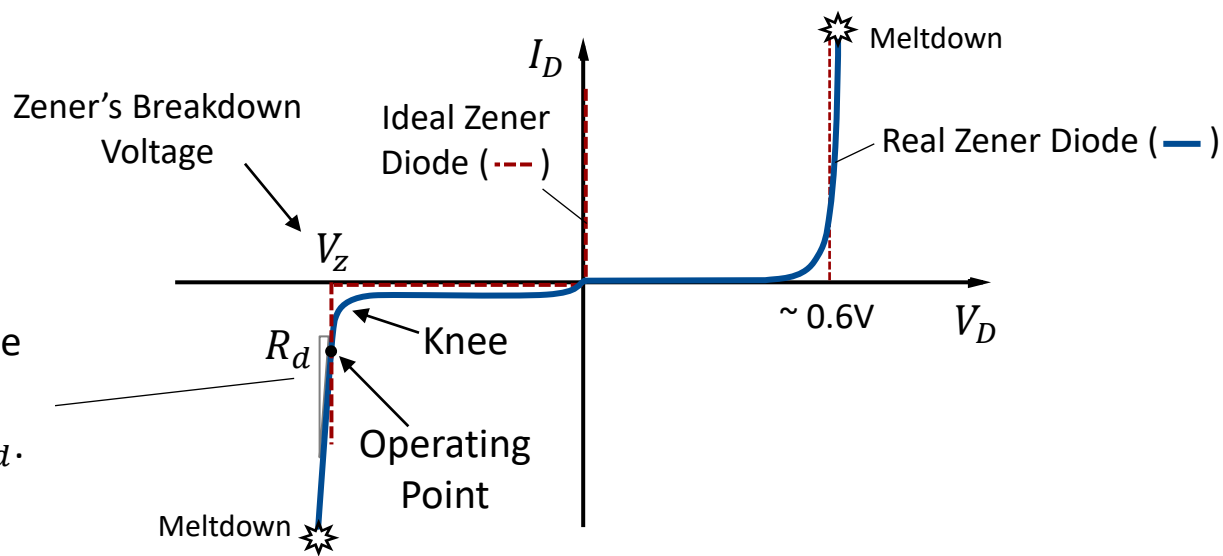
When a diode is reverse biased with a large enough voltage, the diode allows a large reverse current to flow. This is called diode breakdown. For most diodes the breakdown value is at least 50 V. However, a special class of diodes known as **Zener**, **avalanche**, or **voltage-regulator diodes** is designed to exhibit **steep breakdown curves** with **well-defined breakdown voltages** (between 1.8 and 200 V). Thus, they can **maintain** a nearly **constant (DC) voltage** in the presence of a **variable supply voltage and variable load** resistance over a wide range of currents. Hence, they are good candidates for building **simple voltage regulators**.



(Schematic Symbol)

\* The slope of the Zener diode curve at an operating point is called **Dynamic Resistance**  $R_d$ .

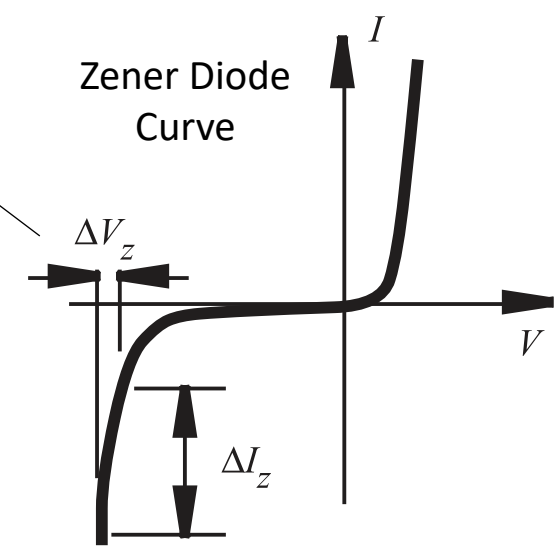
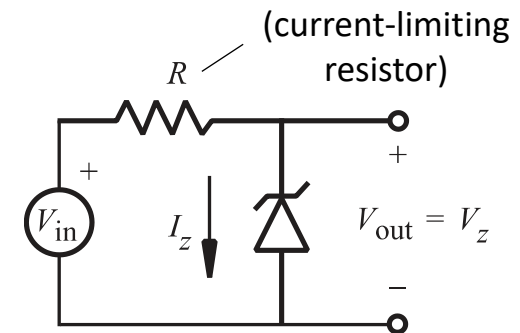
$$R_d = \frac{\Delta V_D}{\Delta I_D}$$





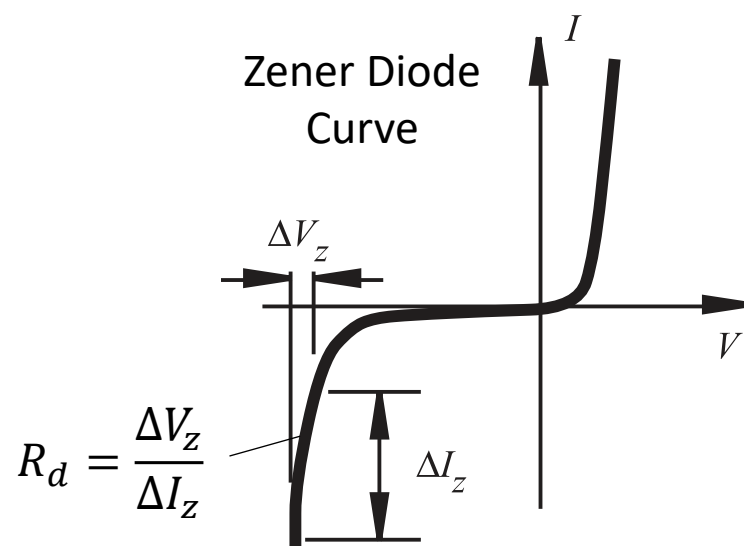
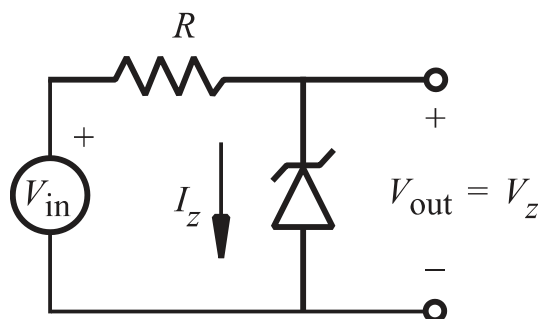
# Zener Diode Voltage Regulator

- To properly use the Zener diode in a circuit, the Zener should be **reverse biased** with a voltage kept in **excess** of its breakdown or Zener voltage  $V_Z$ .
- Using a Zener diode in series with a resistor results in a simple **voltage regulator** in a way that the output voltage  $V_{out}$  of the circuit is maintained or regulated by the Zener diode at the Zener voltage  $V_Z$ .
- When the current through the Zener diode changes ( $\Delta I_Z$ ), output voltage remains **relatively** constant ( $\Delta V_Z$  is small).  $\Delta V_Z$  is a measure of the voltage regulation of the circuit.
- This simple voltage regulator is more effective, only if the input voltage and load do not change much (see the next example).



# Example

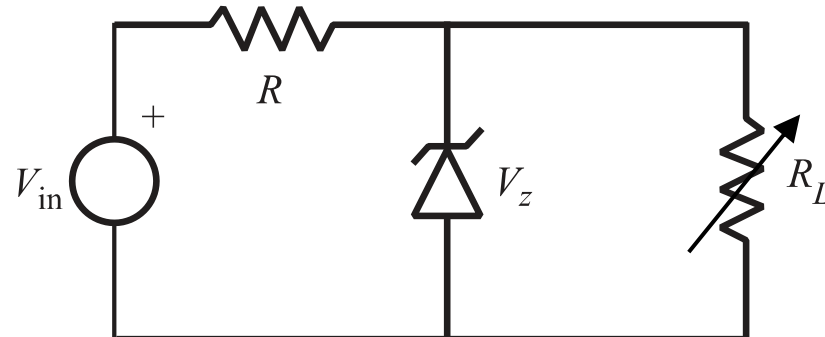
Express changes in the regulator output voltage  $\Delta V_{\text{out}}$  in terms of fluctuations in the source voltage  $\Delta V_{\text{in}}$ .



# Example: Zener Diode Voltage Regulator Design

Suppose we need to design a regulated 15 V DC source to power a mechatronic system, and we would like to use the following voltage regulator circuit. Furthermore, suppose we have access to only a poorly regulated DC source  $V_{in}$  whose nominal value is 24 V (assume that the Zener diode is ideal). Find the minimum required current-limiting resistance  $R$ .

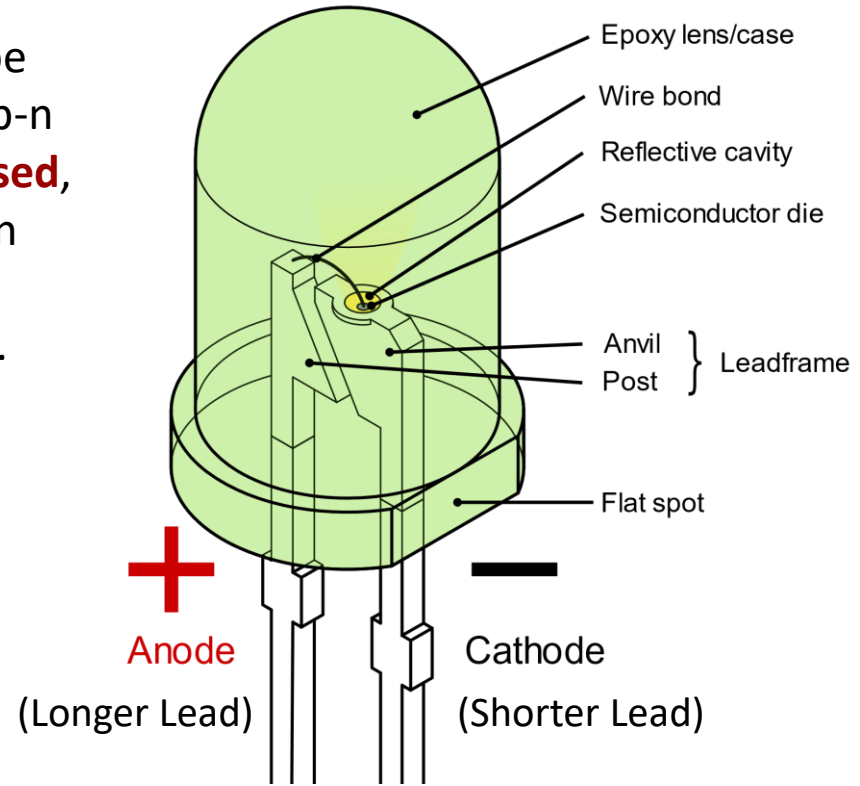
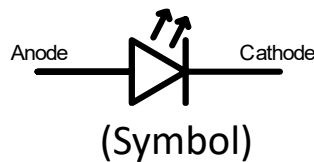
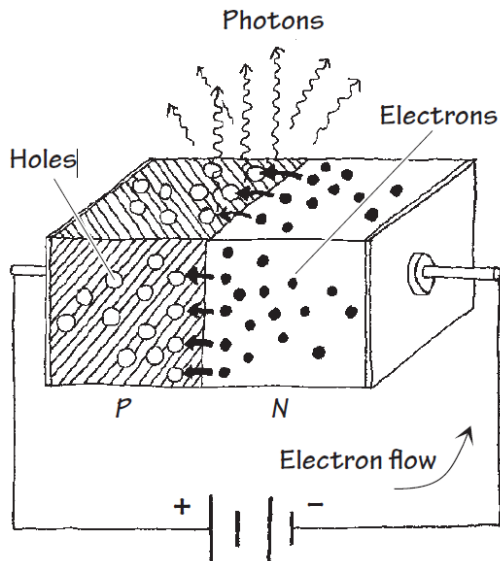
Maximum value of the load  $R_{L_{max}}$  is 240  $\Omega$ , and we want to select a 1 W Zener diode.





# Light-Emitting Diode (LED)

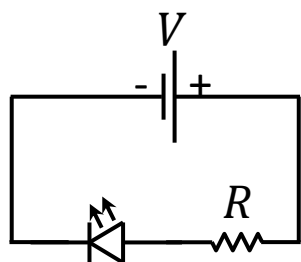
Light-Emitting Diodes are made by joining n-type and p-type semiconductors together to form a p-n junction. When this p-n junction is **forward-biased**, electrons in the n side are **excited** across the p-n junction and move into the p side, where they combine with holes, then, **photons** are emitted.



- LEDs are manufactured to produce a variety of colors.
- They are usually encased in a colored plastic material that enhances the generated wavelength.
- LED has a voltage drop of 1.5 to 2.5 V when forward biased.

# A Practical Example

Find a proper resistor ( $R$ ) which is needed to be wired in series with an LED in the shown circuit when (a)  $V = 5\text{ V}$ , (b)  $V = 9\text{ V}$ . Based on the LED's datasheet, the voltage drops across the LED is designed to be  $2\text{ V}$  and the maximum current designed to go through an LED is  $30\text{ mA}$ .



$$\begin{array}{l}
 \text{(a) } V_{\text{LED}} = 2\text{ V} \quad \rightarrow \quad V_{\text{R}} = 5\text{ V} - 2\text{ V} = 3\text{ V} \\
 I_{\text{LED}} = 30\text{ mA} \quad \rightarrow \quad I_{\text{R}} = 30\text{ mA} \\
 \left\{ \begin{array}{l} R = \frac{V_{\text{R}}}{I_{\text{R}}} = 100\ \Omega \\ P_{\text{R}} = V_{\text{R}} I_{\text{R}} = 90\text{ mW} \end{array} \right. \\
 \\
 \text{(b) } V_{\text{LED}} = 2\text{ V} \quad \rightarrow \quad V_{\text{R}} = 9\text{ V} - 2\text{ V} = 7\text{ V} \\
 I_{\text{LED}} = 30\text{ mA} \quad \rightarrow \quad I_{\text{R}} = 30\text{ mA} \\
 \left\{ \begin{array}{l} R = \frac{V_{\text{R}}}{I_{\text{R}}} = 233.33\ \Omega \\ P_{\text{R}} = V_{\text{R}} I_{\text{R}} = 210\text{ mW} \end{array} \right.
 \end{array}$$

- The **intensity of light** is related to the **amount of current** flowing through the device. However, it is important to include a series **current-limiting resistor** in the circuit to prevent excess forward current, which can quickly destroy the diode.



# Bipolar Junction Transistors (BJTs)





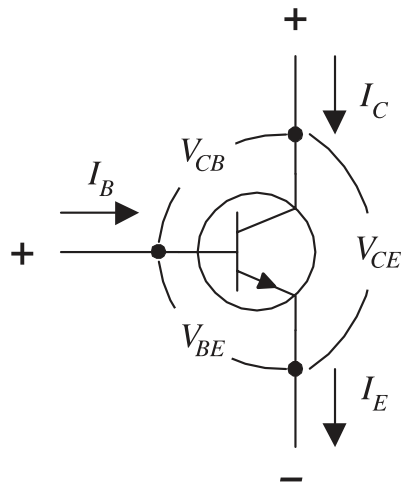






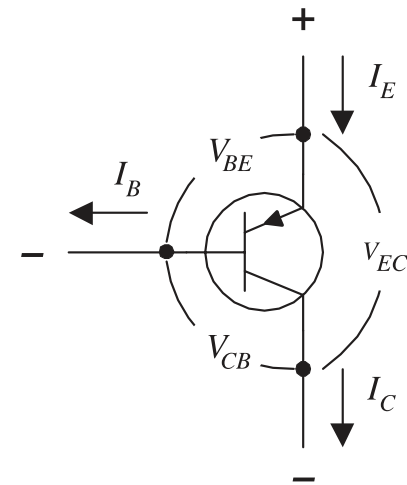
# NPN & PNP BJT

A **npn** bipolar transistor uses a small input current and positive voltage at its base (relative to its emitter) to control a much larger **collector-to-emitter** current.



$$\begin{aligned} V_{BE} &= V_B - V_E \\ V_{CE} &= V_C - V_E \\ V_{CB} &= V_C - V_B \\ I_E &= I_C + I_B \end{aligned}$$

A **pnp** transistor uses a small output base current and negative base voltage (relative its emitter) to control a larger **emitter-to-collector** current.



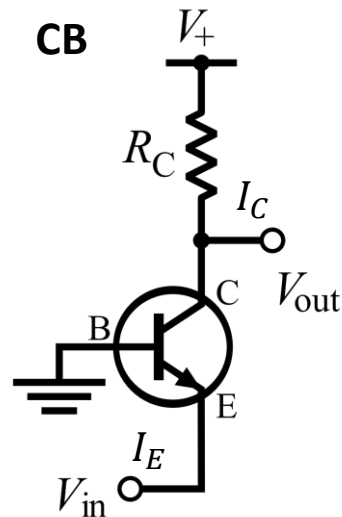
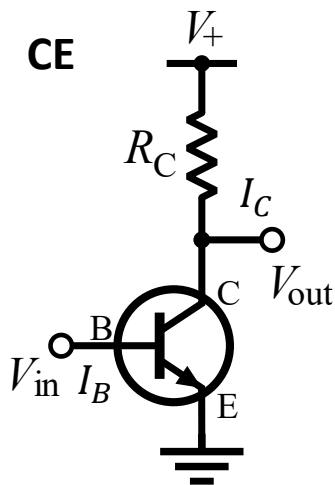
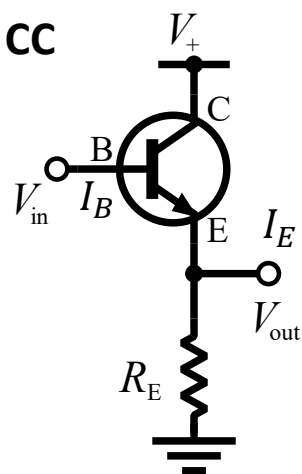
# BJT Configurations

- ❖ As BJT is a **three** terminal device, there are three possible ways to connect it within an electronic circuit with one terminal being **common** to both the **input** and **output**.
- ❖ Each configuration responds differently to its input signal within a circuit.
- ❖ In all configurations, the emitter junction is forward biased and the collector junction is reverse biased.

- **Common Collector (CC) Configuration**

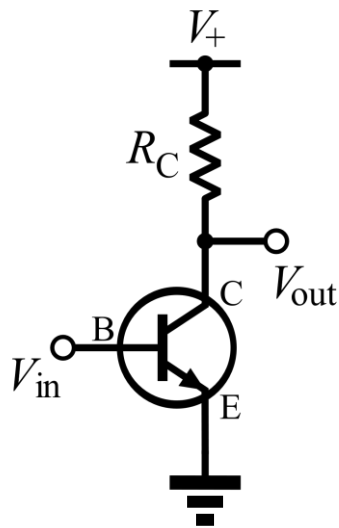
- **Common Emitter (CE) Configuration** ← frequently used

- **Common Base (CB) Configuration**

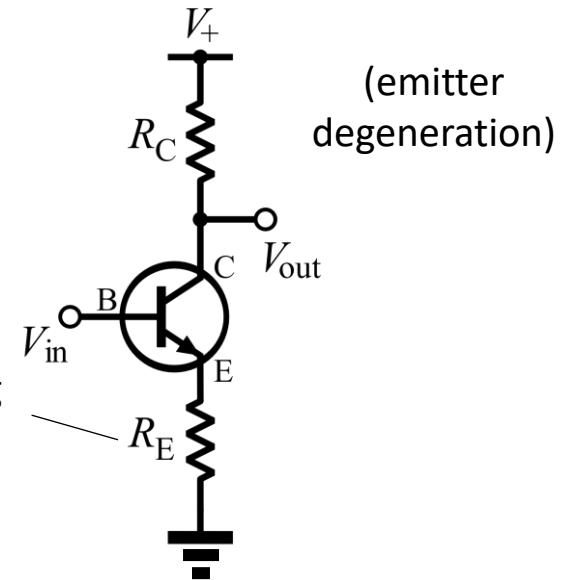


# Common Emitter (CE) Configuration

In this configuration, the **base** terminal of the transistor serves as the **input**, the **collector** is the **output**, and the **emitter** is **common** to both (for example, it may be tied to ground reference or a power supply rail). This type of configuration is the most commonly used circuit for transistor-based amplifiers and switches.



A small resistor for improving the distortion and stability characteristics of the circuit.

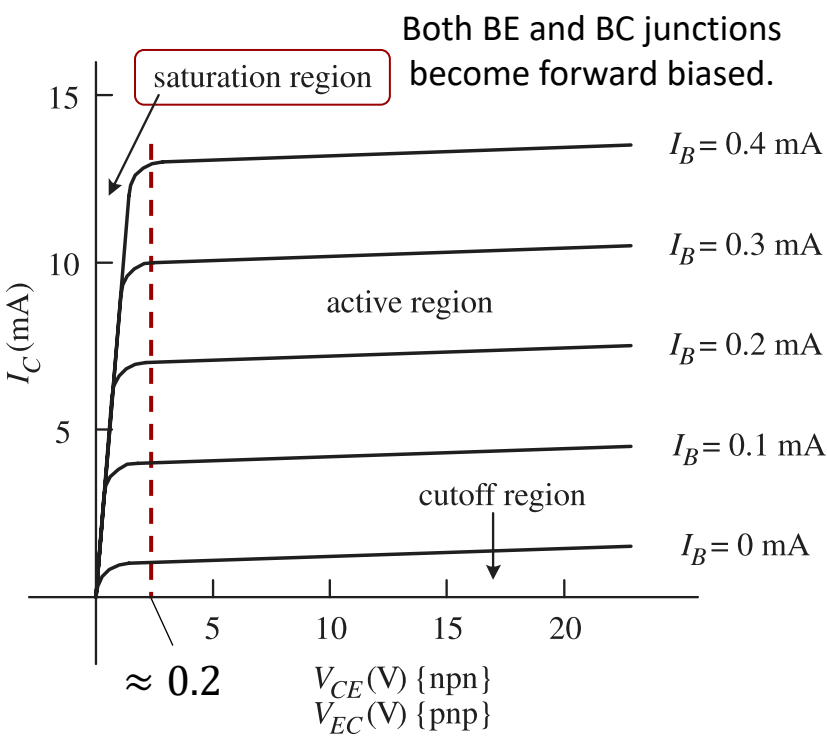


**BJT Operation Modes** are **Cutoff Mode**, **Saturation Mode**, and **Active Mode**.



# BJT Operation Modes: Saturation Mode

When  $V_{BE} \cong 0.7V$ , (and consequently,  $I_B > 0$  (sufficiently)) and  $V_{CE}$  reaches its minimum ( $V_{CE} \leq 0.2$  for a BJT), **maximum** collector current flows to emitter and the transistor acts much like a closed switch (i.e., transistor is **Fully ON**).



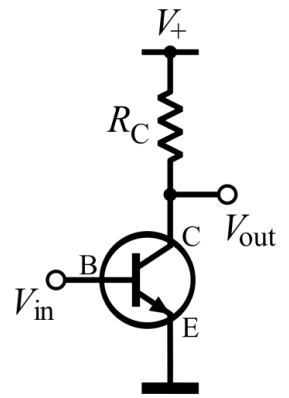
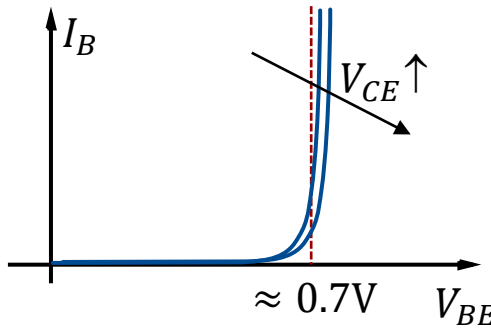
$$I_E = I_C + I_B$$

$$I_C \gg I_B \quad I_C \neq \beta I_B$$

$$V_{BE} \cong 0.7 \text{ V}$$

$$V_{CE} \cong 0.2 \text{ V}$$

$I_C$  is determined by  $R_C$



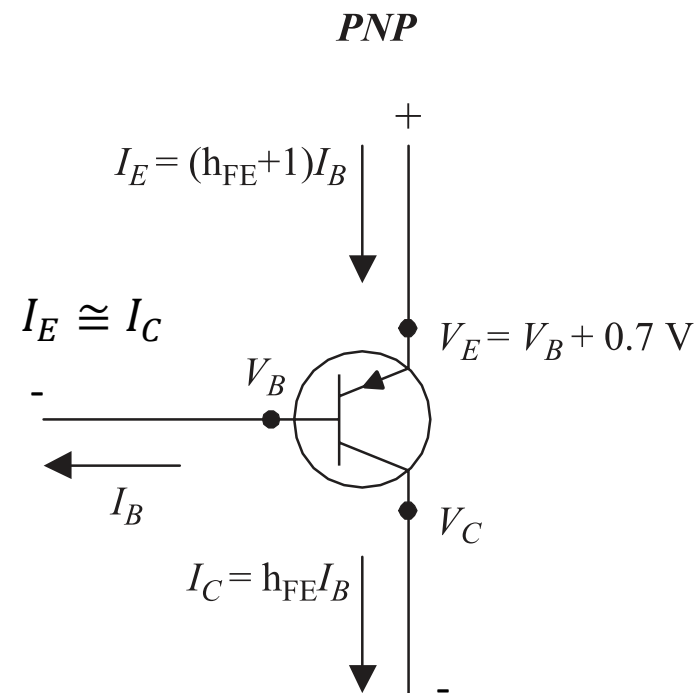
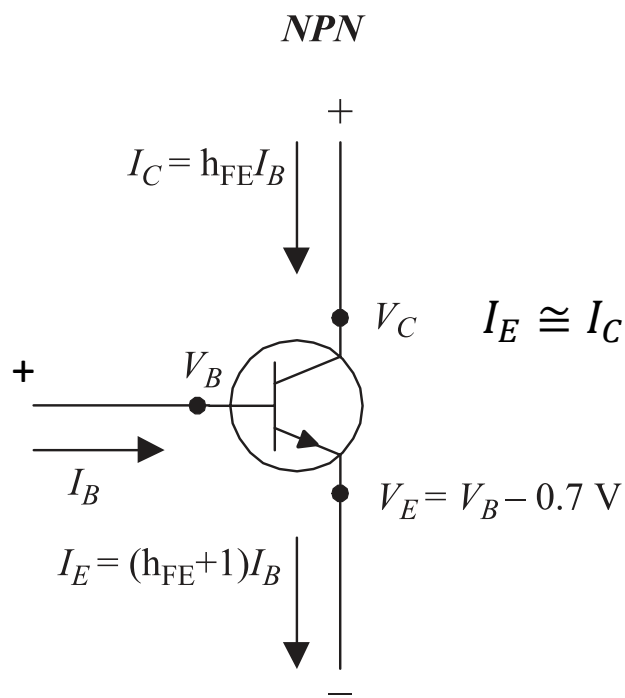
**Note:** In this mode, power dissipation of BJT is min:

$$P_{BJT} = I_C V_{CE}$$

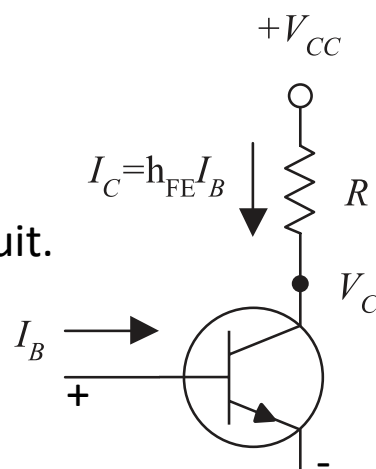
**Note:** In this mode, the  $I_C$  is independent of  $I_B$ , as long as there is enough  $I_B$  to ensure saturation.



# Voltages & Currents in Active Mode



- $V_C$  depends on the circuit.  
For example:



$$V_C = V_{CC} - I_C R$$

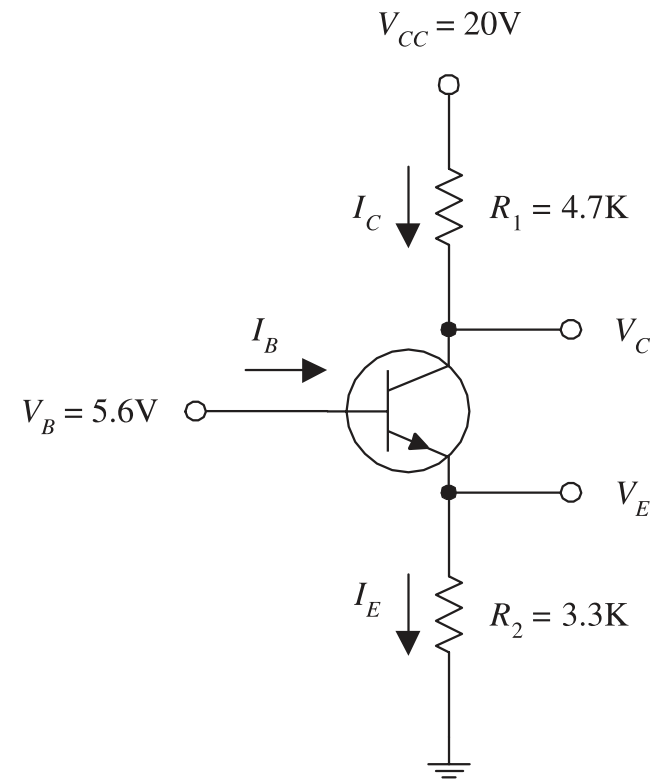
$$V_C = V_{CC} - (h_{FE} I_B) R$$





# Example

Assume that the transistor is in Active Mode and  $\beta = 100$ , find  $V_E$ ,  $I_E$ ,  $I_B$ ,  $I_C$ , and  $V_C$  ( $V_{BE} = 0.6\text{ V}$ ).



# Transistor as a Switch

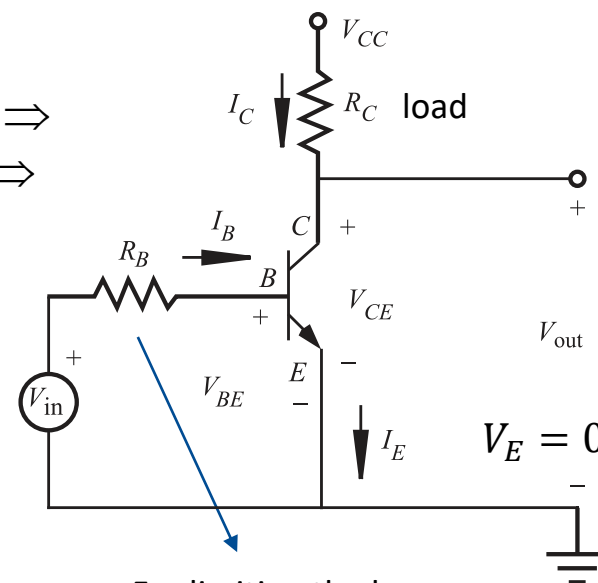
The simplest way to switch moderate to high amounts of power using a low-output current device (e.g., ICs) is to use a transistor. When using a transistor as a **switch** it must be either in **Cutoff Mode** (Fully-OFF) or **Saturation Mode** (Fully-ON). The **Common Emitter (CE)** configuration is more appropriate because it is easy to saturate the transistor.

## Cutoff or Fully-OFF Mode:

$V_{in} < 0.7V \Rightarrow V_{BE} < 0.7V \Rightarrow$  BE junction is reverse biased  $\Rightarrow$   
 $I_B = 0 \Rightarrow I_C = I_E \cong 0$  (leakage)  $\Rightarrow V_{out} = V_{CC}, V_{BE} = V_{in} \Rightarrow$   
 open switch.

## Saturation or Fully-ON Mode:

$V_{in} > 0.7V$  (sufficiently)  $\Rightarrow V_{BE} \cong 0.7V \Rightarrow$  BE junction  
 is forward biased  $\Rightarrow I_B > 0$  (sufficiently)  $\Rightarrow$  BC  
 junction is forward biased  $\Rightarrow V_{CE}$  reaches its minimum  
 ( $V_{CE} = V_{out} \leq 0.2V$  for a BJT)  $\Rightarrow$  maximum  $I_C$  flows  $\Rightarrow$   
 closed switch.



For limiting the base  
 current.

$$I_B = \frac{V_{in} - V_{BE}}{R_B}$$

$$V_{CC} - V_C = R_C I_C$$

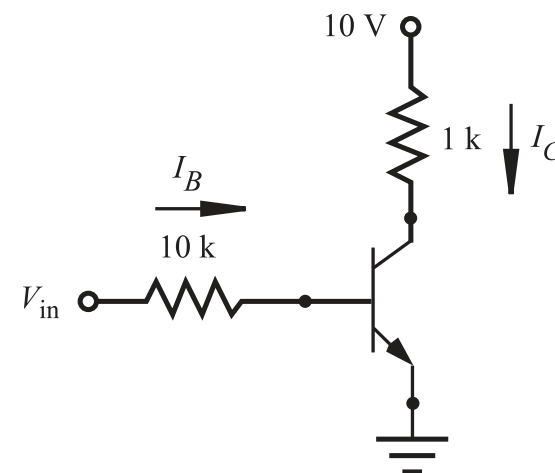
# Example: Guarantee of Saturation Mode

When designing a transistor switch, we need to guarantee that the transistor is fully saturated when it is on.

In the following circuit, what **minimum** input voltage  $V_{in}$  is necessary to saturate the transistor?

Transistor characteristics:

- Maximum collector current (continuous) = 200 mA
- $V_{CE}(\text{sat}) = 0.2 \text{ V}$
- $V_{BE}(\text{sat}) = 0.7 \text{ V}$
- $h_{FE} = \beta = 100$  (depending on collector current and many other things)



**Note:** Normally, we use a voltage larger than calculated minimum  $V_{in}$  (e.g., 2 to 5 times larger) to ensure that the transistor is fully saturated.

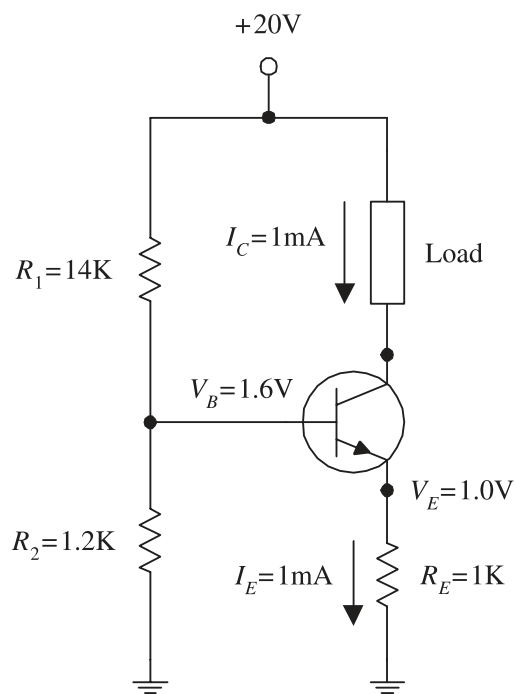




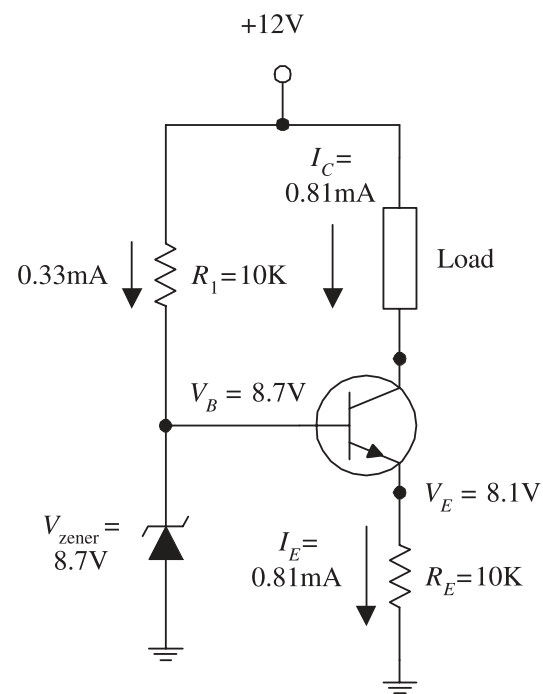
# Application: Current Source

Two common methods for biasing a current source are to use either a voltage-divider circuit or a Zener diode regulator.

voltage-divider: 
$$V_B = \frac{R_2}{R_1 + R_2} V_{CC}$$

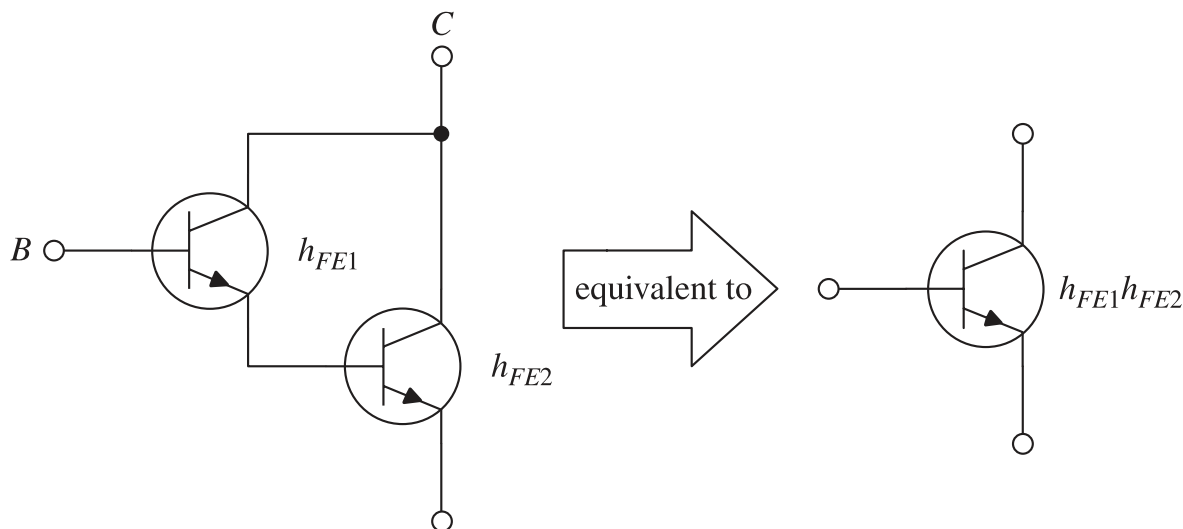


Zener diode regulator: 
$$V_B = V_{\text{zener}}$$



# Darlington Transistor

By attaching two transistors together, a larger  $h_{FE}$  (or  $\beta$ ) equivalent transistor circuit, which is equal to the product of the individual transistor's  $h_{FE}$  values ( $h_{FE} = h_{FE1}h_{FE2}$ ), is formed.



**Darlington Pairs** usually comes in a single package and they are used for **large current** applications and as input stages for amplifiers, where big input impedances are required. They come in npn (D-npn) and pnp (D-pnp) Darlington packages.

